

**CUSTOMER NO. 46850**

**PATENT**

**IN THE UNITED STATES PATENT AND TRADEMARK OFFICE**

**Re: Attorney Docket No. Liu 25-18-17-7**

In re application of: Xiang Liu, Lothar Benedikt Erhard Josef Moeller, Xing Wei, and  
Chongjin Xie

Serial No.: 10/730,413

Filed: 12/08/2003

Matter No.: 990.0506

Group Art Unit:

Examiner:

Phone No.:

2613

Pascal, Leslie C.

571-272-3032

For: Duobinary Receiver

**DECLARATION UNDER 37 CFR 1.132**

Commissioner for Patents  
P.O. Box 1450  
Alexandria, VA 22313-1450

1. I am an applicant of the above-identified patent application ("present application") and an inventor of the subject matter described and claimed therein.
2. I have been working at Lucent Technologies and then Alcatel-Lucent for over ten years, where I am a Member of Technical Staff. Over the years, I have worked on various design aspects of fiberoptic communications systems.
3. The Optical Fiber Communication Conference and Exposition (OFC) is the largest and most highly regarded annual event worldwide in the optical-communications community (see also Exhibit A).
4. OFC postdeadline papers are widely regarded as showing most important and innovative R&D results in the field of fiber optics in each particular year. A jury of about 10-15 top experts from the industry and academia selects a very limited number of postdeadline papers to be presented. Postdeadline-paper sessions are among the best attended at the OFC and most widely quoted (see also Exhibit B).

5. The subject matter of the present application was described in a submission to the OFC held in February of 2004, which submission was accepted and presented thereat as a postdeadline paper entitled "10 Gb/s Duobinary Receiver with a Record Sensitivity of 88 Photons per Bit" (see Exhibit C).

6. The above-indicated submission to the OFC was accepted as a postdeadline paper because said submission demonstrated unexpected, record-setting results. In particular, a record OSNR of about 10.8 dB @BER= $10^{-3}$  had been achieved for optical duobinary signals that were processed at the receiver using a sampling-window width of less than about 25% of the bit length (see the footnote on the fourth page of Exhibit C).

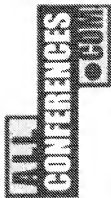
I declare further that all statements made herein of my own knowledge are true and that all statements made on information and belief are believed to be true; and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under Section 1001 of Title 18 of the United States Code and that such willful false statements may jeopardize the validity of the application or any patent issuing thereon.

4/7/09

Date



Lothar Benedikt Erhard Josef Moeller



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## Optical Fiber Communication Conference &amp; Exposition (OFC) 2004

|                        |   |            |                   |            |             |                          |
|------------------------|---|------------|-------------------|------------|-------------|--------------------------|
| Exhibitors             | February 23, 2004   | Ends       | February 27, 2004 | Papers     | 4th         | OFC 2004                 |
| Location               | USA   | State      | California        | City       | Los Angeles | mschup@ofcconference.com |
| Category               | Science   | Category 1 |                   | Category 3 | Exhibitors  | Y                        |
| Engineer/manager       |   |            |                   | Company    |             |                          |
| URL                    | <a href="http://www.ofcconference.org">http://www.ofcconference.org</a>   |            |                   |            |             |                          |
| Website                | <a href="http://www.ofcconference.org">http://www.ofcconference.org</a>   |            |                   |            |             |                          |
| One-point              | The Optical Fiber Communication Conference and Exposition (OFC) is the largest and most highly regarded event in the industry. No other conference draws nearly the audience of buyers, business leaders, executives, managers and developers. That is because only OFC provides both a high-powered, commercial-driven exhibition with leading edge, peer-reviewed educational programming, Plenary Session. Each year, OFC begins with a powerful Plenary Session that addresses the topics, trends and issues at the forefront of the field. |            |                   |            |             |                          |
| Additional information |   |            |                   |            |             |                          |

# Exhibit B

TCPCminutes\_wcnc99\_exhibit.txt

IEEE COMSOC TCPC  
IEEE COMMUNICATIONS SOCIETY  
TECHNICAL COMMITTEE ON PERSONAL COMMUNICATIONS

Minutes of the meeting held at WCNC'99  
Hyatt Regency New Orleans, Tuesday 23 September 1999 (2:00 4:00 pm)

Minutes  
(Minutes taken by Khaled Ben Letaief)

Chair: Zygmunt J. Haas

Present:

Zygmunt Haas, Khaled Ben Letaief, Moe win, Giovanni Corazza, Upkar Varshney, Hamid Aghavami, Demitros Kazakos, Kainam Thomas Wong, Greg Pollini, Jerry Gibson, Dhawal S. Moghe, Mark Haas, Tom Stevenson, Yuguang Fang.

1. The meeting started at 2:10 pm.
2. Minutes of the TCPC meeting held in ICC'99, Vancouver were approved.

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[...]

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## 8. Conferences and Meetings Reports

[...]

5. They discussed a proposal from the Meetings and Conferences Committee for "Post Deadline Papers."

The "post deadline" paper have been used effectively at the OFC conference for approximately a decade to attract last minute submissions on hot topics that might otherwise not be considered. Nim Cheung said that the "post deadline" paper sessions are among the best attended and most widely quoted at OFC.

The process requires the Technical Program Committee to select papers the night before the opening day of the conference based on 3-4 page summaries submitted by potential speakers. At OFC only a very small percentage of papers are typically selected. "Post deadline" paper sessions are typically held from 4-6 in the evening, unlike poster sessions, which often consist of papers that are not strong enough to make the first cut, the post deadline paper submissions tend to be of a high quality and are often submitted by top experts.

[...]

There was no other business. The meeting was adjourned at 3:45 p.m.

*Exhibit C*

**OFc:04**

Optical Fiber Communications  
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# Postdeadline Papers

Technical Conference: February 22-27, 2004

Exposition: February 24-26, 2004

Los Angeles Convention Center  
Los Angeles, California, USA

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Optical Society of America

[www.ofcconference.org](http://www.ofcconference.org)

Received 15 April 1996; accepted 15 May 1996

## 10 Gb/s Duobinary Receiver with a Record Sensitivity of 88 Photons per Bit

Luthar Möller, Chonglin Xie, Roland Ryf, Xiang Liu, Xing Wei

Bell Labs, Lucent Technologies, 901 Holmdel-Keight RD, NJ 07733, USA, [lmoller@lucent.com](mailto:lmoller@lucent.com)

**Abstract:** We demonstrate a novel receiver concept for duobinary signals that allows for data recovery with a sensitivity of  $\sim 19$  dBSm (88 Photons/bit, BER=10<sup>-9</sup>, 1<sup>st</sup> PRBS). Co-sampling together with maximum likelihood sampling phase estimation results in superior back-to-back performance without lowering the large chromatic dispersion tolerance of duobinary signaling.

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OCIS codes: 060.2130; Fiber-optic communications; 060.4680 Modulation

### 1. Introduction

Optical duobinary signaling provides superior protection against chromatic dispersion (CD) impairments when compared to other modulation techniques, making it an interesting candidate for network applications since limits for CD compensation can be reduced <sup>1,2,3</sup>.

However, although the duobinary transmitter technology, similar in complexity to that of its rival, the NRZ signal<sup>4</sup>, has been available for a long time, it has not yet been widely used in commercial products. One reason for this is its weaker back-to-back (b-b) performance. In contrast to published simulations results, predicting that duobinary requires a smaller optical signal to noise ratio (OSNR) than NRZ at a given BER, so far all reported experimental data show the opposite. While compared with NRZ the maximum uncompensated transmission length of duobinary under the assumption of a 2dB (D) penalty can be three times larger, in a b-b performance measurement, it is typically a few dB worse in required OSNR. This fact is of great practical significance since the next generation of TXs must advantageously run on both CD compensated (equivalent to b-b operation) and CD uncompensated links.

Several pulse-shaping techniques, applied on the TX side, have been proposed to overcome the b-b performance problem<sup>5,6</sup>. However, all these techniques seem to improve the b-b performance at the expense of some amount of CD tolerance since they rely on pulse width reduction techniques.

Here we show that timing jitter of the transmitter and the sampling window width of the RX are key design limitations of high sensitivity duobinary transceivers. Based on these insights, we developed a novel RX concept utilizing maximum likelihood optimal sampling phase estimation (MLPE)<sup>7</sup> that allows duobinary reception with a record sensitivity, which is very close to the best reported one for 10Gb/s NRZ<sup>8,9</sup> without losing any CD tolerance.

### 2. Receiver sensitivity enhancement by sampling window width reduction (Quasi Dirac Sampling, QDS)

A duobinary signal with large CD tolerance can be generated by low-pass filtering (LPF) the electrical driver signal, for the Mach-Zehnder modulator. Fig. 1a shows a simulated LPF duobinary b-4 eye diagram, which has a distinct intensity "ripple" on the zero rail and a crossing point higher than that of NRZ<sup>10</sup>. The ripple on the zero rail and the narrow space valley are known to be associated with the good CD tolerance of LPF duobinary signals, but they lead to poor b-b performance of duobinary compared to NRZ. Conventional data decision circuits integrate the signal power over a fraction of the bit length (sampling window) before comparing the integrated signal with a threshold voltage. The undesired optical power near the center of a zero bit inevitably would contribute to the integral and therefore the decision threshold has to be set higher than that of a NRZ signal to minimize the BER. More precisely, if the sampling window height is located too close to the space level (position A, Fig. 1a), the data decision can be corrupted by integrating too much of the energy

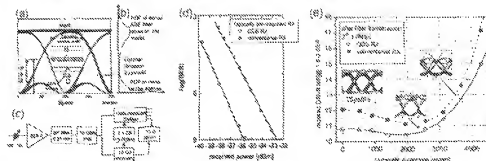


Fig. 1. (a) Schematic of a duobinary receiver, (b) eye diagram of the QDS and ZEPD and the opt. decision threshold, (c) Set-up of the QDS RX, (d) BER curves for the QDS RX and a conventional RX, (e) CD tolerance of the QDS RX and a conventional RX.

# Exhibit C (continued)

stemming from adjacent pulses (ISI). On the other hand, if the window is located too close to the mark level (position B, Fig. 1a), the signal AND beam noise that dominates in an optical pre-amplified RX and is larger on marks, would impair the decision. In other words, the selection of the decision threshold (position C, Fig. 1a) for a conventional RX is a compromise between suffering degradation from ISI and noise and it can not be set to a level that would be the optimum choice regarding the noise distribution function (Fig. 1b) alone in an ISI free signal. Less ISI energy integration and noise collection becomes possible simply by shortening the sampling window width (Quasi Direct Sampling, QDS) and moving the window position deeper in the space valley (position D, Fig. 1a).

## Experimental setup and results

Conventional data decision consists of cascaded master and slave flip-flops that are triggered on the rising edge of the clock tone. To emulate a reduced sampling window width we drive the flip-flops of our customized GaAs ASIC with a clock frequency that is four times higher than the data rate. This yields a window width of approx. 10ps. De-multiplexing of the 40Gbit/s pulse stream recovers the original 10 Gbit/s data pattern. The overall RX architecture is shown in Fig. 1c. Note, the O/E converter (PIN) including the transimpedance amplifier has a bandwidth of approx. equal 80% ( $\sim 80\text{GHz}$ ) the data rate. Thus the noise characteristics of the RX appear to be no worse than those of a conventional RX as one might think due to the higher bandwidth of the data decision circuit.

To demonstrate the performance gain of our RX, we carried out all measurements twice under the same conditions with the exception of using a conventional 10 Gbit/s limiting amplifier (Nemet GB 79), known for its high sensitivity, as a data decision circuit in this case. The commercially available dichotomous TX is based on 1P filtering with 2.5 GHz bandwidth and generates a pp-jitter of less than 25ps.

Fig. 1d shows the BER curves for both RXs obtained when an optical pre-amplifier was used and indicates a QDS RX gain of  $\sim 3\text{ dB}$  at a BER =  $1 \cdot 10^{-9}$  (Sensitivity  $\sim 16\text{dBm}$ ). Since the next generation of transmission systems are expected to run with forward error correction (FEC), we measured the CD tolerance of the RXs at a BER =  $1 \cdot 10^{-9}$ , which is close to the threshold of enhanced FEC with 7% overhead. However, the data rate of the 2<sup>nd</sup> PRBS was kept at 50Gbit/s rate in order to make the results comparable to previously published reports, which mainly focus on 10Gbit/s experiments. The b-b performance of our RX is  $\sim 1.4\text{ dB}$  improved by QDS resulting in  $\sim 10.8\text{ dB}$  required OSNR @ BER =  $10^{-9}$  (measured at 0.1nm OSR resolution). Note, the b-b performance is less than 0.5dB worse compared to that of the optimal amount of CD (at  $\sim 2000\text{ ps/nm}$ ). For a 2 dB OSNR penalty, we measured with the QDS RX tolerable CD amounts of  $\sim 3715\text{ ps/nm}$ .

## 3. Oversampling together with logic signal processing for additional RX sensitivity enhancement

Further RX sensitivity improvement can be achieved when timing jitter mitigation techniques are applied. Fig. 2a shows the simulated eye closures due to jitter for duobinary and regular NRZ. Obviously jitter requires rising the threshold in the case of duobinary to eliminate ISI effects in the decision process, which shifts the threshold away from the optimal value regarding the noise distributions. For the case of NRZ, the same amount of jitter does not require a threshold change due to

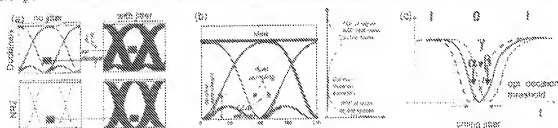


Fig. 2 (a) Threshold lifting due to jitter impairments for duobinary and NRZ. (b) Sketches sampling windows for dual sampling. (c) LO, I+ pattern transitions with jitter relative to the sampling points of dual sampling (alpha) and mono sampling (gamma).

the wide zero valley, for jitter mitigation purpose we extended the QDS RX scheme by establishing more than one sampling point with a short window width per bit slot (see in Fig. 2b the case of dual sampling). Fig. 2c shows the LO, I+ pattern transitions, most susceptible to ISI, after the O/E conversion of the signal when jitter is present. We assume, for the sake of simplicity, that the timing jitter between the adjacent ONE pulses is strongly correlated. Then by patterning dual sampling with sampling points  $\alpha$  and  $\beta$  and combining the obtained data samples through a logical AND operation, a wrong decision on a ZERO made, by i.e.  $\alpha$  can still be overturned by the correct decision of  $\beta$ . A similar argument holds for the case when ASIs corrupts the decision process. It is expected to be unlikely that both pulses are distorted and negatively influence both  $\alpha$  and  $\beta$  since the noise processes in the two neighboring bit slots are only weakly correlated. The impact of signal-ASIs, beat noise generated within the center of the ZERO bit slot (possibly caused by insufficient extinction ratio) and the distinct ZERO ripple of duobinary could be slightly mitigated by oversampling too. In summary, every symbol is sampled more than one time per bit slot. In the case that one sampling result is ZERO, the possibility of different results on neighboring sampling points is eliminated. Since it is more likely that a ZERO than a ONE is corrupted, our procedure maximizes the likelihood that the optimal sampling phase for a ZERO was selected (Maximum